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Multi-level Identification of Hammerstein-Wiener (N-L-N) System in Active Experiment

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 u_k

Context of block-oriented models

 Z_k

 v_k

 $\eta(x,c^*)$



 x_k

 $\left\{\gamma_j^*\right\}_{j=0}^q$

Fundamental blocks:

Linear dynamic (L/LD) Nonlinear static (N/NS)

Relevant structures:

Hammerstein model

Wiener model

Cascade structures

Elementary identification methods:

Least squares method

Kernel estimates



Statement of the problem

SISO Hammerstein-Wiener (N-L-N) system

$$u_{k} \qquad \mu(u, a^{*}) \qquad w_{k} \qquad \{\gamma_{j}^{*}\}_{j=0}^{q} \qquad x_{k} \qquad \eta(x, b^{*}) \qquad v_{k} \qquad y_{k} \qquad y_{k} \qquad y_{k} = \eta \left(\sum_{j=0}^{q} \gamma_{j}^{*} \mu(u_{k-j}, a^{*}), b^{*} \right) + z_{k} \qquad y_{k} = v_{k} + z_{k}, \qquad y_{k} = v_{k} + z_{k}, \qquad v_{k} = \eta(x_{k}), \qquad x_{k} = \sum_{j=0}^{q} \gamma_{j}^{*} w_{k-j}, \qquad w_{k} = \mu(u_{k})$$



Statement of the problem

SISO Hammerstein-Wiener (N-L-N) system

$$\mu(u) = \mu(u, a^*) = a_1^* f_1(u) + a_2^* f_2(u) + \dots + a_m^* f_m(u),$$

$$a^* = (a_1^*, a_2^*, \dots, a_m^*)^T, a^* \in \mathbb{R}^m,$$

 $\eta(x) = \eta(x, b^*) = b_1^* g_1(x) + b_2^* g_2(x) + \dots + b_n^* g_n(x),$

 $b^* = (b_1^*, b_2^*, \dots, b_n^*)^T, b^* \in \mathbb{R}^n$



General assumptions

Regarding the system:

- Linear dynamic block is a FIR filter of known order q.
- Static nonlinear characteristics are Lipschitz functions described by the combination of *a priori* known base functions *f* and *g*. Characteristic $\eta(\cdot)$ is strictly monotonous.

Concerning input and noise signals:

- $u_k^{(1)}$ is an *i.i.d.* random process with Lipschitz probability density function f, and f(0) > 0.
- $u_k^{(2)}$ is an *i.i.d.* random and binary process.
- Input and noise signals are mutually independent and have finite variances. Expected value of noise signal equals zero.



Purpose of identification

 $Q(\gamma, a, b) = E(y_k - \overline{y}_k)^2 \rightarrow \min_{\gamma, a, b}$,

 $y_k = y_k(\gamma^*, a^*, b^*)$ $\overline{y}_k = \overline{y}_k(\gamma, a, b)$



The algorithm

Stages

- Recovery of the finite impulse response γ^*
- Estimation of second nonlinear characteristic (parameters b^*)
- Creation of additional signal r_k with the same expected value as inaccessible signal x_k
- Identification of first nonlinear characteristic (parameters a^{*})

Methods

- Kernel-censored least squares estimate
- Kernel-based method with the help of binary sequences
- Usage of reversed function $\eta^{-1}(\cdot)$ and probability density function f(z)
- Least squares method as in Hammerstein system identification



Identification of linear block

$$\widetilde{\gamma} = \left(\sum_{k=1}^{N} \phi_{k} \phi_{k}^{T} K\left(\frac{\Delta_{k}}{h_{1}}\right)\right)^{-1} \cdot \left(\sum_{k=1}^{N} \phi_{k} y_{k} K\left(\frac{\Delta_{k}}{h_{1}}\right)\right)$$
$$\phi_{k} = \left(u_{k}, u_{k-1}, \dots, u_{k-q}\right)^{T}$$
$$\Delta_{k} = \|\phi_{k}\|_{\infty} = \max_{j=0,1,\dots,q} |u_{k-j}|$$
$$K\left(\frac{\Delta_{k}}{h_{1}}\right) = \begin{cases} 1, as |\Delta_{k}| \leq h_{1}, \\ 0, elsewhere. \end{cases}$$



Identification of linear block

Theorem 1:

Let given assumptions be in force. Then, for the Hammerstein-Wiener system and $h \sim N^{-\alpha}$, where $\alpha \in \left(0, \frac{1}{d}\right)$ and d = q + 3, it holds that $\hat{\gamma}_j \rightarrow \gamma_j^*, j = 0, 1, ..., q$

in probability as $N \to \infty$, provided that $c = \mu'(u_0)\eta'(x_0) \neq 0$.



Identification of linear block

Proof (sketch):

Let's take Taylor series expansion and apply it to input and output nonlinearities around points u_0 and x_0 :

$$\mu(u_k) = \mu(u_0) + c_1(u_k - u_0) + \rho(u_k),$$

$$\eta(x_k) = \eta(x_0) + c_2(x_k - x_0) + \xi(x_k),$$

where $c_1 = \mu'(u_0), c_2 = \eta'(x_0).$

Observations:

 $\rho(u_k), \xi(x_k)$ are second order Taylor's rests and:

 $\begin{aligned} |\rho(u_k)| &= o(h), \\ |\xi(x_k)| &= o(h). \end{aligned}$



Binary representation of number $i - 1, i = 1, 2, ..., N_0, N_0 = 2^{q+1}$

$$d_{1} = (0,0,...,0,0,0)^{T},$$

$$d_{2} = (0,0,...,0,0,1)^{T},$$

$$d_{3} = (0,0,...,0,1,0)^{T},$$

$$\vdots$$

$$d_{N_{0}} = (1,1,...,1,1,1)^{T}.$$

$$x_{[1]}, x_{[2]}, \dots, x_{[i]}, \dots, x_{[N_0]}$$



$$\hat{\eta}(x_{[i]}) = \frac{\sum_{k=1}^{N} y_k \delta(\phi_k, d_i)}{\sum_{k=1}^{N} \delta(\phi_k, d_i)}$$

$$\delta(\phi_k, d_i) = \begin{cases} 1, & as \ \phi_k = d_i \\ 0, & elsewhere \end{cases}$$

$$P\{\delta(\phi_k, d_i) = 1\} = \frac{1}{N_0} = \frac{1}{2^{q+1}}$$



$$y_{[1]}, y_{[2]}, \dots, y_{[L]}, L - \text{random}$$

$$S(x_{[i]}) \triangleq \{y_k \mid \phi_k == d_i\}$$

$$\hat{\eta}(x_{[i]}) = Avg\left(S(x_{[i]})\right) = \begin{cases} \frac{1}{L} \sum_{l=1}^L y_{[l]}, \text{ if } L > 0\\ 0, \text{ otherwise} \end{cases}$$



Theorem 2:

Under give assumptions, it holds that

$$E[\eta(x_{[i]}) - \hat{\eta}(x_{[i]})]^2 \rightarrow 0$$
, as $N \rightarrow \infty$

in each estimation point $x_{[i]} = d_i^T \gamma^*$, $i = 1, 2, ..., N_0$, such that $x_{[i]} \in cont(\eta(), g())$, where $cont(\eta(), g())$ is a set of all point of continuity of $\eta()$ and g().



Proof (sketch):

Using Wald identity and second form of the estimation function:

$$E \hat{\eta}(x_{[i]}) = \eta(x_{[i]}) + E\left(\frac{\sum_{k=1}^{N} (z_k \cdot \delta(\phi_k, d_i))}{\sum_{k=1}^{N} \delta(\phi_k, d_i)}\right) = \eta(x_{[i]}) + \frac{EL \cdot Ez_1}{EL}$$
$$EL = P(\phi_k = d_i) \cdot N = \frac{1}{N_0} \cdot N = \frac{N}{2^{q+1}}$$

and benefitting from conditional variances:

$$var\,\hat{\eta}(x_{[i]}) = \sum_{k=1}^{N} P(L=k) \cdot \frac{\sigma_z^2}{k} = \frac{N_0 \sigma_z^2}{N} = c \cdot \frac{1}{N} \sim N^{-1}$$



$$\left\{ \begin{pmatrix} x_{[i]}, \hat{\eta}(x_{[i]}) \end{pmatrix} \right\}_{i=1}^{N_0} \\ \hat{b} = (\Psi^T \Psi)^{-1} \cdot \Psi^T \zeta \\ \Psi = (\psi_1^T, \psi_2^T, \dots, \psi_{N_0}^T)^T \\ \Psi_i = \left(g_1(x_{[i]}), g_2(x_{[i]}), \dots, g_n(x_{[i]}) \right)^T \\ \zeta = \left(\hat{\eta}(x_{[1]}), \hat{\eta}(x_{[2]}), \dots, \hat{\eta}(x_{[N_0]}) \right)^T$$



Output signal filtration

 r_k

 y_k

ς()

$$y_k = \eta(x_k) + z_k$$

$$x_k = \eta^{-1}(y_k - z_k)$$

$$r_{k} = \varsigma(y_{k})$$

$$\varsigma(y) = E\{x_{k} | y_{k} = y\} = \int_{-\infty}^{\infty} \eta^{-1}(y - z)f(z)dz$$

$$Er_{k} = E\varsigma(y_{k}) = E \int_{-\infty}^{\infty} \eta^{-1}(y_{k} - z)f(z)dz =$$

$$= E \int_{-\infty}^{\infty} \eta^{-1}(\eta(x_{k}))f(z)dz = Ex_{k} \cdot$$

$$\int_{-\infty}^{\infty} f(z)dz = Ex_{k}$$



Identification of first nonlinear characteristic

$$x_k = \phi_k^T \cdot \theta^*$$

$$\lambda_{k} = \left(f_{1}(u_{k}), \dots, f_{m}(u_{k}), f_{1}(u_{k-1}), \dots, f_{m}(u_{k-1}), \dots, f_{1}(u_{k-q}), \dots, f_{m}(u_{k-q})\right)^{T}$$

$$\theta^* = \left(\gamma_0^* a_1^*, \dots, \gamma_0^* a_m^*, \gamma_1^* a_1^*, \dots, \gamma_1^* a_m^*, \dots, \gamma_q^* a_1^*, \dots, \gamma_q^* a_m^*\right)^T$$

$$\widehat{\theta} = (\Lambda_N^T \Lambda_N)^{-1} \cdot \Lambda_N^T R_N$$



Numerical example

$$N = 10^{5}$$

$$u_{k}^{(1)} \sim U[-1,1]$$

$$u_{k}^{(2)} \sim B(0,1)$$

$$z_{k} \sim U[-0.5,0.5]$$

$$\mu(u) = a_{1}^{*}u + a_{2}^{*}u^{2}$$

$$\eta(x) = b_{1}^{*}x + b_{2}^{*}x^{2}$$

$$\gamma^{*} = [0.6 \ 0.3 \ 0.1]^{T}$$

$$a^{*} = [0.8 \ 0.2]^{T}$$

$$b^{*} = [0.7 \ 0.4]^{T}$$



Estimation of parameters a^*

$$E_{\gamma} = \frac{1}{q+1} \sum_{i=0}^{q} (\gamma_i^* - \hat{\gamma}_i)^2 = 1.38 \cdot 10^{-4}$$



0.6

 h_1

0.7

0.8

0.9





Estimation of parameters b^*





Estimation of parameters b*

 $b^* = [0.7 \ 0.4]^T$

 $\hat{b} = [0.699 \ 0.401]^T$





Kernel estimation of f(z)

 $f(z) = \frac{1}{Nh} \sum_{k=1}^{N} K\left(\frac{z_k - z}{h}\right)$ k = 1





Output signal filtration

1) Fast Fourier transform

2) Numerical integration with Riemann sum

3) Monte Carlo method

$$\hat{\zeta}(y) = \frac{1}{n} \sum_{i=1}^{n} \eta^{-1} (y - z_i)$$



Applications

- Modelling both actuator and sensor nonlinearities
- Modeling physical, chemical, biological processes:
 - polymerase reactors
 - ionospheric processes
 - ■pH processes
 - magnetospheric dynamics
- Modelling electrically simulated muscles
- Modelling power amplifiers, electrical drives
- Modelling magneto-rheological dampers



Conclusions

- Combination of both parametric and nonparametric approaches
- The system is identifiable and the solution is unique for the impulse response fulfilling the given assumptions and for output nonlinearity satisfying Haar condition
- Division of the identification into stages significantly reduces dimensionality of the system identification
- Effectiveness is strictly related to the length of the impulse response
- Recommended for dynamic filters with short memory and more elaborated nonlinearities
- The algorithm for H-W system identification can adapt itself both to Hammerstein and Wiener systems separately without any additional knowledge about the examined system



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